

DYNAMIC CCD LINE LENGTH TIMING FOR COLOR SCANNERS

FIELD OF THE INVENTION

5 This invention relates generally to the operation of linear charge coupled device (CCD) imagers and, more particularly, to synchronizing the CCD readout with the motion of the object being scanned.

BACKGROUND OF THE INVENTION

10 When operating a linear film scanner it is important that the film being scanned has constant velocity. In addition, if a CCD imager with electronic exposure control is used, the exposure for every line should be the same. (This technique is shown in commonly assigned U.S. Patent No. 5,105,264, entitled "Color image sensor having an optimum exposure time for each color" and issued April 14, 1992 in the name of H. Erhardt.)

15 In a color film scanner with a tri-linear CCD sensor (i.e., a CCD having three color sensors respectively sensitive to three colors, e.g., red, green and blue), any speed variation in the transport of the film can result in color fringing effects in the resulting image. This effect will be noticed in the slow scan direction and can result in unacceptable image artifacts.

20 In U. S. Patent No. 4,205,337, the time interval between successive scans is varied in order to switch between film formats including television standards of fifty or sixty frames per second, and motion picture film rates of twenty-four and twenty-five frames per second. A variable frequency oscillator is used to set up a chosen operating speed for a given frame rate. While such
25 techniques as disclosed in this patent can effectively handle varying frame rates, what is needed is a technique for line by line compensation of the time interval between successive scans within a frame, and in particular a technique for doing this without creating dark signal artifacts between line captures because of line variability.

SUMMARY OF THE INVENTION

The present invention is directed to overcoming one or more of the problems set forth above. Briefly summarized, according to one aspect of the present invention, the invention overcomes the limitations of conventional systems by utilizing a line clocking arrangement in a scanner for synchronizing the line readout of a clocked imaging device with the motion of an object being scanned. The imaging device includes a photosensitive area, an overflow drain, a horizontal output register, a drain gate interposed between the photosensitive area and the overflow drain, and one or more transfer gates interposed between the photosensitive area and the horizontal output register for transferring charge between the photosensitive area and the horizontal output register. The line clocking arrangement includes an encoder for sensing movement of the object being scanned and generating a sync signal in correspondence with a movement of the object, and a timing generation circuit for generating clock signals for controlling the clocked imaging device. The clock signals include a drain clock signal for controlling the dumping of charge into the overflow drain and an output clock signal for clocking image charge through the horizontal output register. The timing generation circuit receives the sync signal and times the duration of the drain clock signal and the beginning of the output clock signal to the occurrence of the sync signal, whereby the line readout time is dynamically adjusted to changes in velocity of the scanned object during a period when charge is being dumped into the overflow drain.

By using an encoder to monitor the position of the film in the gate, the resulting information can be used to slightly lengthen or shorten the CCD line readout time, thereby synchronizing the line readout to the motion of the scanned object. Of particular advantage, the change in line readout time occurs when the CCD is not integrating electrons but is dumping electrons into the overflow drain. Application of this technique thereby minimizes dark signal differences between line captures.

These and other aspects, objects, features and advantages of the present invention will be more clearly understood and appreciated from a review of the following detailed description of the preferred embodiments and appended claims, and by reference to the accompanying drawings.

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BRIEF DESCRIPTION OF THE DRAWINGS

FIG. 1 is a diagram of a single channel of a linear CCD imager known in the prior art.

FIG. 2 shows a set of waveform diagrams for the line clocking of the linear CCD imager shown in Figure 1.

FIG. 3 shows the waveform diagrams of Figure 2 modified in accordance with the invention to dynamically control the line length of the linear CCD imager shown in Figure 1.

FIG. 4 is a waveform diagram expanding the beginning of the first line shown in Figure 3.

FIG. 5 is a waveform diagram expanding the beginning of the second line shown in Figure 3.

FIG. 6 is a diagram of a single channel of another type of linear CCD imager known in the prior art.

FIG. 7 shows waveform diagrams modified in accordance with the invention to dynamically control the line length of the linear CCD imager shown in Figure 6.

FIG. 8 is a waveform diagram expanding the beginning of the first line shown in Figure 7.

FIG. 9 is a waveform diagram expanding the beginning of the second line shown in Figure 7.

FIG. 10 is a block diagram of a color scanner utilizing dynamic control of the line length of its linear CCD imager in accordance with the invention.

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DETAILED DESCRIPTION OF THE INVENTION

Because CCD imaging devices and line clocking techniques for use therewith are well known, the present description will be directed in particular to elements forming part of, or cooperating more directly with, apparatus in accordance with the present invention. Elements not specifically shown or described herein may be selected from those known in the art.

Referring first to Figure 10, a color scanner is shown to include a light source 1, a circle to line converter 2 for generating a scanning line 2a on an object 3 that is to be scanned, and a lens 4 for imaging the scanning line 2a on a tri-linear CCD imager 5. The tri-linear CCD imager 5 includes separate CCD arrays 5a, 5b and 5c responsive to red, green and blue light, respectively. In the preferred embodiment, the object to be scanned is a transparent color film, such as a photographic negative or positive (slide) film, that is moved in a direction 3a by a conventional film drive mechanism 3b. However, this embodiment is understood to be without limitation, and the scanned object could be something else, such as a print or copy material, and the collection optics could be adapted for a reflective system, instead of the transmissive system as shown in Figure 10. The output signals from the tri-linear CCD imager 5 are digitized in an A/D converter 6 and applied to a digital signal processor 7 for subsequent processing, which is not part of this invention. A synchronizing signal (Sync) is generated by an encoder 8 that indicates the position of the film 3 as it passes through the scanner. The Sync signal is applied to a CCD and datapath timing generation circuit 9, which provides the circuit elements for generating line clocking signals that dynamically control the line length of the linear CCD imager in accordance with the invention.

Figure 1 shows a single channel of a linear CCD imager used in the tri-linear CCD imager 5 shown in Figure 1. The imager includes a photodiode array 10 (composed of individual photodiodes 10a, 10b....), a drain structure (LS drain) 12 and a horizontal shift register 14. A gate (LOGn gates) 16 is interposed between the photodiode array 10 and the drain structure 12, and transfer gates

(TG1 and TG2) 18 are interposed between the photodiode array 10 and the horizontal shift register 14. Horizontal register clocks (PH1 and PH2) 20 are provided to clock the image signal out of the horizontal shift register 14. Output circuitry 22 provides conventional treatment of the output image signal, such as correlated double sampling to provide analog output samples to the A/D converter 6.

Exposure control is implemented by selectively clocking the LOGn gates 16 during portions of the scanning line time. First, the charge collected in the photodiode array 10 is drawn off to the LS drain 12 by positively biasing the LOGn gates 16 while TG1 is in an off state. By changing the LOGn biasing to the opposite state during the line, the charge generated in the photodiodes 10a, 10b... is no longer dumped into the LS drain 12 but instead accumulated in the photodiode array 10. At the end of the exposure period the collected charge is moved into the horizontal shift register 14 using the TG clocks 18 and then read out of the CCD using the Phi clocks 20.

Figure 2 shows the normal clocking operation of the linear imager shown in Figure 1. The line period is denoted by t_{line} in the figure, with LOGn starting the line switching high and TG1 starting the line switching low. With TG1 low and LOGn high the charge collected in the photodiode will be drawn off to the LS drain. At some time later (t_{dump}) the LOGn clock is switched low and the charge is collected in the photodiode. The effective exposure time (t_{exp}) is the net time between the falling edge of the LOGn clock and the falling edge of the TG1 clock.

Figure 3 shows the clocking for dynamically controlling the line length of the of the CCD imager according to the invention. The signal Sync is received by the CCD and datapath timing generation circuit 9 from the encoder 8 that indicates the position of the item to be scanned. Since the velocity of the item to be scanned may have moved faster or slower than nominally expected, the timing generation logic will use this Sync signal to compensate for the variation in velocity. In this example, for the first line capture the item to be scanned has

moved quickly, so the Sync signal has a positive edge transition soon after the
 TG1 falling edge (t_{wait1}) but after the TG2 falling edge. The Sync rising edge is
 used to reset the internal counters in the timing generation logic used for clocking
 the next line of pixels, and will stay high until the TG1 signal switches low at the
 5 end of line (t_{exp_max}). The first line shown in Figure 3 therefor has a minimum line
 length for this first line acquired (t_{line_min}). The item being scanned has not moved
 as quickly for the second line scanned and therefor the length of the second line is
 longer (t_{line_longer}). Notice that the Sync rising edge has been delayed from the
 falling edge of TG1 (t_{wait2} is greater than t_{wait1}) for the second line. In order to
 10 compensate for this increase in line length the time between the falling edge of
 TG2 and the start of the Phi2 clock has been increased as shown in Figure 3. For
 clarity it should be noted that for both lines in this figure the duration of the
 TG1 and TG2 pulses are always the same. In particular is important to hold TG2
 low when waiting for the Sync signal rising edge so as to minimize the generation
 15 of dark signal in the TG2 gate area. Also note that t_{exp} and t_{exp_max} are the same for
 both lines shown, while the t_{line_min} has a shorter duration than t_{line_longer} . The
 advantage of clocking in this way is that the dark signal generated during the time
 when TG2 switches low and the start of electronic exposure when LOGn goes low
 will be dumped into the LS drain 12 of the CCD imager. The result is that both
 20 lines captured will have the same integrated dark signal from the photodiode,
 which is the major contributor to dark signal in the CCD.

Figure 4 expands the beginning of the first line shown in Figure 3
 while Figure 5 expands the beginning of the second line shown in Figure 3. In
 these figures the TG1 (t_{pd}) and the TG2 ($t_{pd} + t_{tg1}$) pulsewidths are the same. The
 25 section of the line that varies is the time between the falling edge of TG2 and the
 start of the Phi2 clock (t_{tg2_min} versus t_{tg2_longer}). As a result, the amount of time
 spent dumping charge into the LS drain 12 is extended from t_{dump} to t_{dump_longer} . In
 both Figure 4 and Figure 5, the time between the rising edge of Sync and the start
 of the Phi2 clock is constant (t_{sync_Phi}). Note that in this example the LOGn signal

goes low and charge is acquired in both Figure 4 and Figure 5 at pixel 3 of the line, resulting in the same t_{exp} time in both cases.

Figure 6 shows a single channel of a second known example of a linear CCD imager that has a single TG gate with an adjacent accumulation region. The imager includes a photodiode array 30 (composed of individual photodiodes 30a, 30b...), a overflow drain structure (LS drain) 32, a pair of accumulation regions 34a and 34b and a pair (odd and even) of horizontal shift registers 36a and 36b. A gate structure (LOGn gates) 37 is interposed between the accumulation regions 34a and 34b and the drain structure 32, and a single transfer gate connection (TG) 38 is interposed between the accumulation regions 34a and 34b and the horizontal shift registers 36a and 36b. Respective horizontal register clocks (PH1 and PH2) 40a and 40b are provided to clock the image signal out of the horizontal shift registers 36a and 36b. Output circuitry 42 provides conventional treatment of the output image signal, such as correlated double sampling to provide analog output samples to the A/D converter 6. In operation, the charge moves immediately out of the photodiodes 30a, 30b... and into the accumulation regions 34a and 34b, where it can be dumped to the overflow drain structure 32 or transferred to the horizontal registers 36a and 36b, depending on the state of the gate structure 37 and the transfer gate connection 38. The LS and PhiA gates are set with a DC bias while the TG and LOGn gates are clocked for vertical transfer.

Figure 7 shows the line clocking of a CCD imager that is shown in Figure 6. As before, the signal Sync is received by the CCD and datapath timing generation circuit 9 from the encoder 8 that indicates the position of the item to be scanned. Since the velocity of the item to be scanned may have moved faster or slower than nominally expected, the timing generation logic will use this Sync signal to compensate for the variation in velocity. The first line in this figure has a minimum line time (t_{line_min}) while the second line has a longer line time (t_{line_longer}). Notice that the total electronic exposure time is the same for both lines (t_{exp}).

Figure 8 expands the beginning of the first line shown in Figure 7 while Figure 9 expands the beginning of the second line shown in Figure 7. Once again it is critical that the extra line time be located between the TG falling edge and the start of the Phi2 clock. In these figures the TG (t_{pd}) pulsewidth is the same for both the minimum and longer lines. Notice that the LOGn has a minimum dump duration of t_{dump_min} , so the Sync pulsewidth must have at least a t_{dr_min} minimum time in the low state. The advantage of clocking in this way is once again that the dark signal generated in the photodiode during the time when TG switches low and the start of electronic exposure when LOGn switches low will be dumped into the LS drain 37 of the CCD imager.

The invention has been described in detail with particular reference to certain preferred embodiments thereof, but it will be understood that variations and modifications can be effected within the spirit and scope of the invention.

PARTS LIST

- 1 light Source
- 2 circle to line converter
- 2a scanning line
- 3 film
- 3a direction of film motion
- 3b film drive mechanism
- 4 lens
- 5 tri-linear CCD sensor
- 5a red CCD
- 5b green CCD
- 5c blue CCD
- 6 A/D converter(s)
- 7 digital Signal Processing
- 8 encoder
- 9 CCD and datapath timing generation
- 10 photodiode array
- 10a,b individual photosite
- 12 drain structure
- 14 horizontal shift register
- 16 gate
- 18 transfer gates
- 20 horizontal register clocks
- 22 output circuitry
- 30 photodiode array
- 32 overflow drain structure
- 34a accumulation region
- 34b accumulation region
- 36a odd horizontal shift register
- 36b even horizontal shift register

- 37 gate structure
- 38 transfer gate connection
- 40a horizontal register clock
- 40b horizontal register clock
- 42 output circuitry

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